# A Novel Solution based on Multi-Objective AI Techniques for Optimization of CMOS LC\_VCOs

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Abstract— A method of optimizing components and transistors sizing for CMOS Cross-Coupled LC voltage controlled oscillators is presented in this paper. The design constrains of power consumption, phase noise, and the Figure of Merit (FoM) of LC\_VCOs are applied on Multi-Objective AI techniques, simultaneously. The design parameters of LC\_VCOs are obtained from the two strong algorithms, the Multi-Objective Inclined Planes system Optimization (MOIPO) and the Multi-Objective Particle Swarm Optimization (MOPSO). It was implemented in MATLAB, to the Pareto Optimal Front (POF) solutions, which have an amazingly trade-off between three objective functions. The LC VCO circuits were simulated using this method in a 0.18µm-CMOS process by HSPICE RF environment. The results show that the size of components of the PMOS-only and NMOS-only integrated LC\_VCOs and the optimal trade-off curve between minimum power and minimum phase noise.

*Index Terms*— Artificial intelligence techniques, cross-coupled LC\_VCOs, Multi-Objective Inclined Planes system Optimization (MOIPO), Multi-Objective Particle Swarm Optimization (MOPSO).

## I. INTRODUCTION

Optimization is a mature technology that has been studied extensively by researchers over the last half century [1]. Over the years, optimization methods have evolved considerably and many algorithms and implementations are now available and used in the engineering optimization community. Artificial Intelligence (AI) techniques, such as neural network models, fuzzy logic, expert systems, and heuristic and evolutionary algorithms, are expected to play important roles in solving complex engineering designs. The Voltage-Controlled Oscillator (VCO) is one of the most important blocks in RF communication systems. It is still a challenge to achieve a good performance with low phase noise, low power consumption and wide frequency tuning [2]. There has to be a trade-off among those key requirements and finally, the optimal FoM. VCOs are vital components of PLL circuits, which play an indispensable role in Radio Frequency (RF) Integrated Circuits (IC) for wireless communications like Bluetooth or Wireless Local Area Network (WLAN) [3].

In the design of oscillator circuits, designers often aim at both minimum phase noise and minimum power consumption for a given oscillation frequency. Since, these two objectives are contradictory, this tradeoff is reflected in the Figure-of-Merit (FoM) as follows, where  $\omega_0$  is the oscillation frequency, *P* is the DC power consumption,  $\Delta \omega$  is the offset from the output frequency and  $L(\Delta \omega)$  is the oscillator phase noise. Given by following equation, the original lesson's equation, where, *Q* is the loaded quality factor of the oscillator, *k* is the Boltzmann's constant, *T* is the absolute temperature,  $P_{sig}$  is the oscillation output power, *F* is the noise factor of the amplifier and  $\Delta \omega_1/f_3$  is the corner frequency between  $\omega_1/f_2$  and  $\omega_1/f_3$  portion of the phase noise spectrum [3].

$$FoM = L(\Delta\omega) + 10\log\left[\frac{P}{1mW} \left(\frac{\Delta\omega}{\omega_0}\right)^2\right]$$
(1)

$$L(\Delta\omega) = 10\log\left\{\frac{2FkT}{P_{sig}}\left[1 + \left(\frac{\Delta\omega}{2Q\Delta\omega}\right)^2\right]\left(1 + \frac{\Delta\omega_1/f_3}{|\Delta\omega|}\right)\right\}$$
(2)

Different LC VCO design methodologies have been proposed, either to achieve low phase noise [2], or low power [4], or targeting low FOM [5]. But as discussed in [6], there are several challenges in RFIC design in improving the important circuit parameters, i.e. power, noise, linearity, gain, power supply and frequency. In LC VCOs there are only some imprecise formulas as the rule of thumb for calculation of phase noise and power consumption. Therefore, the design of an optimized LC VCO needs many time-consuming, and perhaps blind, trials and errors. Utilizing Multi-Objective evolutionary techniques such as Inclined Planes system Optimization (MOIPO) and Particle Swarm Optimization (MOPSO) algorithm can turn these blind attempts into an organized series of iterations directed at obtaining power, phase noise, and FoM-optimized designs for LC VCOs (e.g. NMOS-only and PMOS-only). The multi-objective algorithms MOIPO and MOPSO can produce pareto fronts and trade-off diagrams. Phase noise, power consumption, and FoM as the three major objectives are used in the form of a unique technique. HSPICE RF and HBOSC analysis are used for simulation. There are other optimization algorithms that were used to optimize RF circuit [2], [7]-[12]. Furthermore, in the three SA, IPO, and PSO, the objectives were merged into a unique objective using a weighting technique.

In the next section, the cross-coupled LC\_VCOs that was chosen as an application example for the proposed technique is presented. Section 3 presents a case study of the proposed tool/technique and a brief description of the algorithms used. Section 4 details the constraints and performance/fitness functions models that were taken into account in the proposed optimization technique. Section 5 presents the simulation results and compares the performance of LC\_VCOs. Finally, the conclusions are offered in section 6.

# II. LC\_VCO DESCRIPTION

During the design, two VCO architectures were taken into consideration (see Fig. 1). When dealing with high frequency applications, LC\_VCO circuits are preferred to other proposed oscillator structures, such as the ring oscillators [7]. Different structures of LC oscillators have been already studied. In fact, the CMOS cross-coupled LC\_VCO offers better rise- and fall-time symmetry, which results in a smaller  $1/f^{\circ}$  noise corner. In addition, the VCO bias current must be doubled for the NMOS structure to obtain the same tank amplitude as in CMOS structure [7].



Figure 1: a) The PMOS-only cross coupled LC\_VCO, b) NMOS-only cross coupled LC\_VCO

## III. ARTIFICIAL INTELLIGENCE TECHNIQUES FOR OPTIMIZATION OF CMOS LC\_VCOS

In the past few decades, there has been a widespread interaction between researchers seeking for various AI methods to determine the best solutions for a given function. The AI techniques are developed by mimicking or simulating the processes found in nature.

#### A. Multi-Objective Optimization

Pareto Optimality: Multi-Objective Optimization problems consist of simultaneously optimizing several objective functions [13]. It considers, without the loss of generality, the minimization of the *n* components  $F_k$ , k=1,...,n of a vector function *F* of a vector variable *x* in a search space  $\varphi \subseteq R^m$ ,

with

$$F(X) = [F_1(x), \dots, F_n(x)]^T$$
(3)

It is assumed that all the  $F_i$  take on real values so that ordering is possible. If the different objectives are competitive, there is no meaningful solution to this problem because it is not possible to evaluate the relevance of each objective with respect to the others. The most accepted concept of optimality for multi-objective problems is the pareto optimality which is defined as follows:

**Pareto Optimality**. A given  $x^* \in \varphi$  is said to be pareto

optimal or non-inferior if there is no other  $x \in \varphi$  such that

$$\forall i \in \{1, \dots, n\}, \tag{4}$$

$$E_i(x^*) \land \exists i \in \{1, \dots, n\} \not\models E_i(x) \leq E_i(x^*) \tag{5}$$

 $F_i(x) \le F_i(x^*) \land \exists j \in \{1, \dots, n\} | F_j(x) \le F_j(x^*)$ (5) For a given multi-objective problem, the pareto-optimal set

*P* is the set of all the  $x \in \varphi$  that are pareto-optimal and the

pareto front  $P^{f}$  is the set of all the optimal vector function values [13].

### B. Multi-Objective Particle Swarm Optimization (MOPSO) Algorithm

Particle Swarm Optimization (PSO) [14] is a stochastic meta-heuristic created to optimize nonlinear functions based on the movement of bird flocks looking for food. In this method, a swarm (population) of particles (solutions) moves across the search space (evolve) guided by personal and social leaders. To expand a PSO algorithm into a MOPSO, two main modifications are usually done: the creation of an external repository to store the non-dominated solutions, and the use of a leader selection method to select a global leader for the particles among a set of equally good solutions according to some criterion. When the external repository becomes full, an archiving strategy is needed to prune it and keep it on a predefined size, discarding some non-dominated solutions based on some criterion. This criterion has great impact in the quality of the solutions generated in the search, especially in many-objectives due to the large portion of the population that becomes non-dominated. There are many approaches in the literature to manage the repository, and a comparison among some of them is done in [15]. Additionally, two of the approaches that prune the repository and presented good results are the Ideal and Multilevel Grid Archiving (MGA) archives [15].

Since there is no single optimal solution in MOPs, a leader selection method is also needed, and this method has impact on the quality of the solutions as well. A comparison between some of the leader selection methods available in the literature is presented in [15], and three of them that had good results are the Crowding Distance (CD) and NWSum method [15]. Another aspect that has been observed in a MOPSO is that in some conditions the velocity of the particles can become too high, generating erratic movements towards the limits of the decision space. To avoid such situations, the Speedconstrained Multi-objective PSO (SMPSO) algorithm presents a velocity constriction mechanism based on a factor x that varies based on the values of the influence factor of the social  $(C_1)$  and personal  $(C_2)$  leaders. In SMPSO, the crowding distance metric is used in both, the leader selection method and the archiving strategy. Due to its good results in the literature, SMPSO is an algorithm frequently used as a reference [15].

# C. Multi-Objective Inclined Planes system Optimization (MOIPO) Algorithm

The method of IPO is designed based on the dynamic of sliding motion along a frictionless inclined surface [16]. In this algorithm each agent, named "tiny ball" (similar to the particles in PSO) searching the problem space to find the optimum solution. In nature and on surface of the earth, when an object (here, tiny ball) is elevated, it loses potential energy and automatically goes to lower elevation levels. These phenomena in physic are named as the gravitational force which applies to any objects by the earth.

In IPO algorithm, each ball has 3 specifications: position (x), height (h) and angels  $(\phi)$  in relation to the other balls. Ball's positions are feasible solutions for the problem and fitness function is used to calculate the height of each ball. For a clear explanation, we assume a system by N ball as the following:

$$x_i^{\min} \le x_i \le x_i^{\max}, \quad 1 \le i \le N_d \tag{6}$$

where  $x_i$  is a decision variable, *i* is the dimension and  $N_d$  is the number of dimension. An example of search space in IPO algorithm is shown in Fig. 2 with 3 balls. The purpose of IPO is to find the minimum of fitness function  $f(x_1, x_2, ..., x_{Nd})$  which is defined on the problem space. The angle between the *i-th* ball and *j-th* one at interval *t* is calculated in the following equations where,  $f_j(t)$  is the value of fitness function (height) for the *i-th* ball in time *t*.



Figure 2: An example of search space with 3 balls (in IPO)

$$\Phi_{ij}^{d}(t) = \left( \tan^{-1} \left( \frac{f_{j}(t) - f_{i}(t)}{x_{i}^{d}(t) - x_{j}^{d}(t)} \right) \right),$$
(7)  
for  $d = 1, ..., n$  and  $i, j = 1, 2, ..., N, i \neq j$ 

The acceleration amount and direction are calculated using the following equations:

$$a_{i}^{d}(t) = \sum_{j=1}^{N} U(f_{j}(t) - f_{i}(t)) . \sin(\phi_{ij}^{d}(t)),$$
(8)

where U(.) is the unit step function:

$$U(w) = \begin{cases} 1 & w > 0 \\ 0 & w \le \end{cases}$$
(9)

In each iteration of IPO algorithm, the position of each ball is updated by using the following equations, in which  $k_1$  and  $k_2$  are two changing constants with time,  $rand_1$  and  $rand_2$  are two random variables in the range [0 1] and  $v_i^d(t)$  is *i-th* ball velocity in dimension *d* in time *t*.  $x_{best}$  is the best ball position until current iteration.

$$x_i^d(t+1) = k_1 \cdot rand_1 \cdot a_i^d(t) \cdot \Delta t^2 + k_2 \cdot rand_2 \cdot v_i^d(t) \cdot \Delta t + x_i^d(t), \quad (10)$$

$$v_i^d(t) = \frac{x_{best}^d(t) - x_i^d(t)}{\Delta t},$$
(11)

$$k_1(t) = \frac{c_1}{(1 + \exp((t - shift_1) \times scale_1))},$$
(12)

$$k_2(t) = \frac{c_2}{(1 + \exp((t - shift_2) \times scale_2))},$$
(13)

In the above equations  $c_1$ ,  $c_2$ , *shift*<sub>1</sub>, *shift*<sub>2</sub>, *scale*<sub>1</sub> and *scale*<sub>2</sub> are experimentally determined constants for each function [10], [16].

In MOIPO, the algorithm of pareto optimality method is used to identify non-dominant position and it uses an external repository for maintaining the positions [17]. The initial population is randomly generated according to the specified range, then the fitness of population is calculated and eventually the best balls in an external repository (Pareto Solutions) are maintained.

Next according to the IPO algorithm, the position of the ball for the next iteration of the algorithm is updated. This update includes putting all the balls that are not currently dominant in the tank. Simultaneously, each ball that has been dominated by the process of the tank can be removed. Each time the tank of limitations has passed, and considering that the storage capacity is limited, it is imputed proportional to the number of balls in each of hypercube to every possible hypercube and by the roulette wheel, hypercube selected and randomly surplus points will be removed. This process continues until the storage capacity reaches to a quorum [17].

#### IV. THE PROPOSED TECHNIQUE

Using the proposed optimization tool, several LC\_VCOs with different fosc were optimized and simulated. In this paper, we use two multi-objective meta heuristic called the MOPSO and MOIPO. These algorithms are based on a communication tool between the HSPICE/RF and MATLAB software. In short, this tool works as follows. After generating the circuit's netlist, the parameter's values are introduced in this netlist. Then, the MATLAB calls HSPICE/RF was used to check the constraints and evaluate the performances. These performances are then introduced in the optimization algorithm and 'new' parameter's values are generated, etc [17]. Next, each configuration of the LC\_VCOs is optimized as depicted in Fig. 1. So, each one is designed for wide oscillation frequency in a 0.18µm-CMOS technology, in which the following conditions are satisfied while the phase noise, power consumption, and FoM, are used in the form of a unique technique, which are optimized simultaneously. The user-defined constraints are the size of transistors, inductors, bias voltage (V<sub>bias</sub>), capacitors and quality factor of inductors. The design considerations can be stated as follows:

Optimize = phase noise & power consumption & FoM

The user-defined constrains are listed in Table 1.

Table 1 The User-Defined Constraints

Design Considerations	Parameter	Unit	Value
oscillation frequency	$f_{osc}$	GHz	1.1-2.3
width of transistors	$W_{n,p,tail}$	μm	10-100
length of transistors/technology	L	μm	0.1-0.2
bias voltage	$V_{bias}$	V	0.55 - 1
inductors	LI	nH	1-9
capacitors	CI	pF	1 - 7
quality factor of inductors	Q	-	7-13
power supply	$V_{DD}$	V	1.8

The proposed optimization tool optimized each structure and determined the value of design variables, transistors sizes, tank inductances and capacitances and bias voltage ( $V_{bias}$ ) of tail transistor. For simplification of implementation, in circuit (Fig. 1) of any two sorts of cross-coupled LC\_VCOs, inductors are connected to resistor as serial and Q inductors were considered with this resistor (Fig. 3). Capacitances were considered pure. MOSFETs were considered with capacitances and resistors parasitic using parasitic models in radio frequency.

$$L \qquad Q_{S} = \frac{L \otimes_{0}}{R_{S}} \qquad L_{S} \qquad R_{S}$$

Figure 3: Inductance model

When MOIPO and MOPSO are applied as two optimizer core in MATLAB, any ball or particle is included with a value of design variables. Some parameters of MOIPO and MOPSO algorithms are shown in Table 2 and 3.

Table 2 Some Parameters of MOIPO Algorithm

More				Paramete	ers		
MOIPO	iter.	balls	Nrep	$C_{I}$	<i>C</i> <sub>2</sub>	scale <sub>1</sub>	$scale_2$
Value	20	17	100	6.3	2.3	0.71	0.93
MOIDO				Paramete	ers		
MOIPO	sh	ift I	shift2	alpha	Ngrid	beta	gamma
Value	1	0.1	14.1	0.1	10	4	2

Table 3 Some Parameters of MOPSO Algorithm

	Parameters						
MOPSO	iteration	particles	Nrep	Cl	C2	W	
Value	20	17	100	3	2	1	
	Parameters						
MOPSO	Wdamp	alpha		Ngrid	beta	gamma	
Value	0.95	0.1		10	4	2	

# V. RESULTS

The design of two sorts of cross-coupled LC\_VCOs is performed using of mathworks MATLAB R2013b. Simulation temperature was a default temperature, (i.e. 25C). Implementation is carried out with MSI CX41, Intel (R) Core (TM) i5-3230M CPU @ 2.60GHz processors, 6GB RAM. The fitness functions and the limitations are examined by HSPICE A-2010.03-SP1, simulation software. Using the introduced multi-objective evolutionary algorithms, the trade-off curves (Pareto Front) are shown in Figure 4 to 7 for each of the algorithms, for NMOS-only and PMOS-only LC\_VCOs, are obtained. Note that only trade-off for both objectives phase noise and power have been drawn (due to the conflict between them).



Figure 4: Resulting pareto front based on MOIPO algorithm, PMOS-only LC VCO







The simulation results of phase noise using the HBOSC method with HSPICE RF are shown in Figure 8 to 11.

Figure 8: The simulated phase noise using the HBOSC method based on MOIPO algorithm, PMOS-only LC\_VCO



Figure 9: The simulated phase noise using the HBOSC method based on MOIPO algorithm, NMOS-only LC\_VCO



Figure 10: The simulated phase noise using the HBOSC method based on MOPSO algorithm, PMOS-only LC\_VCO



Figure 11: The simulated phase noise using the HBOSC method based on MOPSO algorithm, NMOS-only LC\_VCO

Each point on pareto front presents an optimized design. Therefore, the designer can select each configuration, and extract the circuit parameters for his/her desirable design. The extracted data for each configuration are presented in Table 4 to 7.

Table 6
The Extracted Data for each Configuration (Pareto front Solutions) based on
MOIPO Algorithm for PMOS-only LC_VCO

Table 4
The Extracted Data for each Configuration (Pareto front Solutions) based on
MOIPO Algorithm for NMOS-only LC_VCO

Pareto			Parame	eters	
front	Ln,tail	$W_n$	Wtail	Ll	C1
Solutions↓	(µm)	(µm)	(µm)	(nH)	(pF)
1	0.1528	52.59	20.94	4.129	10
2	0.1530	63.41	26.36	3.603	1.412
3	0.1786	60.55	26.35	3.555	1.753
4	0.1589	63.72	26.46	3.452	1.361
Pareto			Parame	eters	
front	$V_{bias}$	$f_{osc}$	power	phase noise	FoM
Solutions↓	(mV)	(GHz)	(mW)	(dBc/Hz)	(dBc/Hz)
1	650	2.16	1.3	-125.56	191.14
2	650	1.98	1.6	-129.11	192.94
3	650	1.83	1.4	-127.29	191.26
4	650	2.05	1.6	-128.14	192.4

Pareto	Parameters					
front Solutions↓	$L_{p,tail}$ ( $\mu m$ )	$W_p$ ( $\mu m$ )	W <sub>tail</sub> (μm)	L1 (nH)	C1 (pF)	
1	0.1797	10	19.88	4.972	1.875	
2	0.1735	10	10	4.991	1	
3	0.1794	10	10.02	4.969	1.218	
4	0.1686	10	10.26	5	1	
	Parameters					
Pareto			Parame	eters		
Pareto front Solutions↓	V <sub>bias</sub> (mV)	f <sub>osc</sub> (GHz)	Parame power (mW)	eters phase noise (dBc/Hz)	FoM (dBc/Hz)	
Pareto front Solutions↓	V <sub>bias</sub> (mV) 749.13	f <sub>osc</sub> (GHz) 1.62	Parame power (mW) 2.3	phase noise (dBc/Hz) -139.89	FoM (dBc/Hz) 200.56	
Pareto front Solutions↓ 1 2	V <sub>bias</sub> (mV) 749.13 108.21	fosc (GHz) 1.62 2.198	Parame power (mW) 2.3 1.2	eters phase noise (dBc/Hz) -139.89 -136.44	FoM (dBc/Hz) 200.56 202.42	
Pareto front Solutions↓ 1 2 3	V <sub>bias</sub> (mV) 749.13 108.21 758.46	fosc (GHz) 1.62 2.198 2.009	Parame power (mW) 2.3 1.2 1.2	eters phase noise (dBc/Hz) -139.89 -136.44 -128.42	<i>FoM</i> ( <i>dBc/Hz</i> ) 200.56 202.42 193.86	
Pareto front Solutions↓ 1 2 3 4	V <sub>bias</sub> (mV) 749.13 108.21 758.46 818.94	fosc (GHz) 1.62 2.198 2.009 2.204	Parame power (mW) 2.3 1.2 1.2 1	ters phase noise (dBc/Hz) -139.89 -136.44 -128.42 -127.54	<i>FoM</i> ( <i>dBc/Hz</i> ) 200.56 202.42 193.86 194.22	

Table 7 The Extracted Data for each Configuration (Pareto front Solutions) based on MOPSO Algorithm for PMOS-only LC\_VCO

Pareto			Parame	eters		
front Solutions↓	$L_{p,tail}$ ( $\mu m$ )	$W_p$ ( $\mu m$ )	W <sub>tail</sub> (µm)	L1 (nH)	C1 (pF)	
1	0.159	51.15	44.93	2.327	2.426	
2	0.157	56.53	43.10	2.038	2.346	
3	0.149	59.75	42.86	2.415	2.750	
4	0.156	54.90	44.96	1.930	2.345	
Pareto		Parameters				
front Solutions↓	V <sub>bias</sub> (mV)	f <sub>osc</sub> (GHz)	power (mW)	phase noise (dBc/Hz)	FoM (dBc/Hz)	
front Solutions↓ 1	V <sub>bias</sub> (mV) 870.99	fosc (GHz) 2.02	power (mW) 4	phase noise (dBc/Hz) -134.99	FoM (dBc/Hz) 195.1	
front Solutions↓ 1 2	V <sub>bias</sub> (mV) 870.99 900	fosc (GHz) 2.02 2.18	power (mW) 4 3.4	phase noise (dBc/Hz) -134.99 -132.32	FoM (dBc/Hz) 195.1 193.8	
front Solutions↓ 1 2 3	V <sub>bias</sub> (mV) 870.99 900 863.45	fosc (GHz) 2.02 2.18 2.25	power (mW) 4 3.4 3.9	phase noise (dBc/Hz) -134.99 -132.32 -132.85	FoM (dBc/Hz) 195.1 193.8 194.0	

The results of the comparison between performances obtained from using the proposed technique are presented in Figure 4 to 11 and Tables 4 to 7, and some of the published papers [3], [10], [12], [18]-[20]. It is obviously clear that new

Table 5
The Extracted Data for each Configuration (Pareto front Solutions) based on
MOPSO Algorithm for NMOS-only LC VCO

Pareto			Parame	eters	
front Solutions↓	L <sub>n,tail</sub> (µm)	$W_n$ ( $\mu m$ )	W <sub>tail</sub> (µm)	L1 (nH)	C1 (pF)
1	0.1	56.63	20.55	2.838	1.728
2	0.11	64.58	28.45	3.077	1.494
3	0.111	48.58	20	2.985	1.633
4	0.1	55.60	20	3.106	1.319
Pareto			Parame	eters	
front Solutions↓	$V_{bias}$ (mV)	f <sub>osc</sub> (GHz)	power (mW)	phase noise (dBc/Hz)	FoM (dBc/Hz)
1	745.44	2.105	3.6	-127.76	188.64
2	721.73	2.109	4.1	-129.44	189.75
3	783.36	2.114	4.2	-129.93	190.17
4	752.46	2.254	4.2	-129.70	190.37

MOIPO algorithm and MOPSO algorithm could reach to the optimum results for the design of cross-coupled LC\_VCOs. Furthermore, the new MOIPO algorithm shows better results than the PSO algorithm in optimization of two sorts Cross-Coupled LC\_VCOs. Therefore, by using an appropriate multi-objective approach, the global minimal is explored and shown to the designer. They represent a significant design trade-off between the phase noise and power consumption with values of power consumption ranging from 1 to 4.2 mW, phase noise from -139.89 to -125.56 dBc/Hz and FoM from 188.64 to 202.42 dBc/Hz, which are a value well within the wireless applications.

#### VI. CONCLUSION

In this paper, an optimization tool based on two multiobjective evolutionary algorithms, Multi-Objective Inclined Planes system Optimization (MOIPO) and Multi-Objective Particle Swarm Optimization (MOPSO), were presented and applied to the design and optimization of two sorts crosscoupled LC VCOs in which, the power, phase noise and Figure of Merit (FoM) of the LC VCOs are minimized simultaneously. The optimized results include a center frequency within 2.1GHz, the medium 2.6mW power consumption, the medium -130.7dBc/Hz phase noise at 1MHz offset, and medium 193.4dBc/Hz FoM. The comparison of results show a promising potential of the proposed tool for wireless and RF applications due to their capability to generate multiple solutions along the Pareto Optimal Front (POF) as well as their efficiency in parallel execution. For future work, the robustness of the proposed algorithm could be tested with transistors in different process corners and then further analysed the performance with statistical analysis.

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